Indian Institute of ScienceE9-252: Mathematical Methods and Techniques in Signal Processing<br/>Instructor: Shayan G. Srinivasa<br/>Homework #3 Solutions, Fall 2017<br/>Solutions prepared by Priya J NadkarniLate submission policy: Points scored = Correct points scored  $\times e^{-d}$ , d = # days late<br/>Assigned date: Sept. 11<sup>th</sup> 2017

## PROBLEM 1:

Solve problems 4.6 and 4.11 from P. P. Vaidyanathan's book. (8 + 10 points)Solution:

(4.6) For the first system, we obtain,

$$y_{0}(n) = h_{k}(n) * y(n) \Rightarrow Y_{0}(z) = H_{k}(z)Y(z)$$

$$h_{k}(n) = h_{0}(n)\cos\frac{2\pi kn}{L} = \frac{h_{0}(n)e^{\frac{j2\pi kn}{L}} + h_{0}(n)e^{\frac{-j2\pi kn}{L}}}{2}$$

$$H_{k}(z) = \frac{H_{0}(e^{\frac{-j2\pi k}{L}}z) + H_{0}(e^{\frac{j2\pi k}{L}}z)}{2}$$

$$Y_{0}(z) = \frac{H_{0}(e^{\frac{-j2\pi k}{L}}z)Y(z) + H_{0}(e^{\frac{j2\pi k}{L}}z)Y(z)}{2}$$

For the second system,  $y_1(n) = (h_0(n) * y(n)) \cos \frac{2\pi kn}{L}$ . Let  $x(n) = h_0(n) * y(n)$ , then,

$$X(z) = H_0(z)Y(z)$$

$$Y_1(z) = \frac{X(e^{\frac{-j2\pi k}{L}}z) + X(e^{\frac{j2\pi k}{L}}z)}{2}$$

$$= \frac{H_0(e^{\frac{-j2\pi k}{L}}z)Y(e^{\frac{-j2\pi k}{L}}z) + H_0(e^{\frac{j2\pi k}{L}}z)Y(e^{\frac{j2\pi k}{L}}z)}{2}$$

 $Y_0(z) = Y_1(z)$  only if  $Y(e^{\frac{-j2\pi k}{L}}z) = Y(z)$ . Otherwise,  $Y_1(z) \neq Y_0(z)$ . One such case where they are equal is when  $k \mod L = 0$ . Another example where it is equal is when the signal has been upsampled by L (Refer Prob 4.5). Thus, they are not necessarily the same. They may be same in some cases, but differ in others.

(4.11) Consider the following system:

$$x(n) \xrightarrow{\qquad \qquad } H(z) \xrightarrow{\qquad \qquad } x_1(n) \xrightarrow{\qquad \qquad } x_2(n) \xrightarrow{\qquad \qquad } H(z) \xrightarrow{\qquad \qquad } x_3(n) \xrightarrow{\qquad \qquad } s(n) \xrightarrow{\qquad \qquad } y(n)$$

When we pass x(n) through the filter H(z), we obtain

$$X_1(z) = X(z)H(z)$$

When we downsample  $x_1(n)$  by 2, we obtain,

$$X_2(z) = \frac{X_1(z) + X_1(-z)}{2}$$

When we pass  $x_2(n)$  through the filter H(z), we obtain,

$$X_3(z) = X_2(z)H(z)$$

When we downsample  $x_3(n)$  by 2, we obtain,

$$S(z) = \frac{X_3(z) + X_3(-z)}{2}$$

When we upsample s(n) by 4, we obtain,

$$Y(z) = S(z^4)$$



PROBLEM 2:

Consider the following system:

$$x(n) \longrightarrow 3 \longrightarrow H(z) \longrightarrow y(n)$$



Suppose the spectrum of the original signal and transfer function is as given above.

Analyze the spectrum of y(t). Analyze the output spectrum if the decimator and expander and interchanged. (4+3 points)

Solution: Let  $x_1(n)$  and  $x_2(n)$  denote the intermediate signals.



If the decimator and the interpolator are exchanged,

$$\begin{aligned} X_1(z) &= X(z^3) \\ X_2(z) &= \frac{1}{3} \sum_{k=0}^2 X_1(z^{1/3} e^{-j\frac{2\pi k}{3}}) \\ &= \frac{1}{3} \sum_{k=0}^2 X((z^{1/3} e^{-j\frac{2\pi k}{3}})^3) = \frac{1}{3} \sum_{k=0}^2 X(z) = X(z) \\ Y(z) &= 3X(z) \end{aligned}$$

The last equality follows as X(z) is bandlimited to  $\frac{\pi}{3}$ .